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MICROCHIP MCP4901/4911/4921

8/10/12-Bit Voltage Output Digital-to-Analog Converter with SPI Interface

Features

- MCP4901: 8-Bit Voltage Output DAC
- MCP4911: 10-Bit Voltage Output DAC
- MCP4921: 12-Bit Voltage Output DAC
- Rail-to-Rail Output
- SPI Interface with 20 MHz Clock Support
- Simultaneous Latching of the DAC Output with $\overline{\text{LDAC}}$ Pin
- Fast Settling Time of 4.5 μs
- Selectable Unity or 2x Gain Output
- External Voltage Reference Input
- External Multiplier Mode
- 2.7V to 5.5V Single-Supply Operation
- Extended Temperature Range: -40°C to $+125^{\circ}\text{C}$

Applications

- Set Point or Offset Trimming
- Precision Selectable Voltage Reference
- Motor Control Feedback Loop
- Digitally-Controlled Multiplier/Divider
- Calibration of Optical Communication Devices

Related Products

P/N	DAC Resolution	No. of Channels	Voltage Reference (V_{REF})
MCP4801	8	1	Internal (2.048V)
MCP4811	10	1	
MCP4821	12	1	
MCP4802	8	2	
MCP4812	10	2	
MCP4822	12	2	
MCP4901	8	1	External
MCP4911	10	1	
MCP4921	12	1	
MCP4902	8	2	
MCP4912	10	2	
MCP4922	12	2	

Note: The products listed here have similar AC/DC performances.

Description

The MCP4901/4911/4921 devices are single channel 8-bit, 10-bit and 12-bit buffered voltage output Digital-to-Analog Converters (DACs), respectively. The devices operate from a single 2.7V to 5.5V supply with an SPI compatible Serial Peripheral Interface. The user can configure the full-scale range of the device to be V_{REF} or $2 \cdot V_{\text{REF}}$ by setting the gain selection option bit (gain of 1 of 2).

The user can shut down the device by setting the Configuration Register bit. In Shutdown mode, most of the internal circuits are turned off for power savings, and the output amplifier is configured to present a known high resistance output load (500 k Ω , typical).

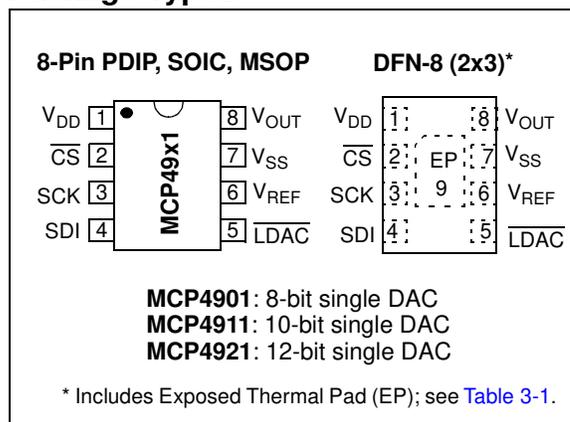
The devices include double-buffered registers, allowing synchronous updates of the DAC output using the $\overline{\text{LDAC}}$ pin. These devices also incorporate a Power-on Reset (POR) circuit to ensure reliable power-up.

The devices utilize a resistive string architecture, with its inherent advantages of low Differential Non-Linearity (DNL) error and fast settling time. These devices are specified over the extended temperature range ($+125^{\circ}\text{C}$).

The devices provide high accuracy and low noise performance for consumer and industrial applications where calibration or compensation of signals (such as temperature, pressure and humidity) are required.

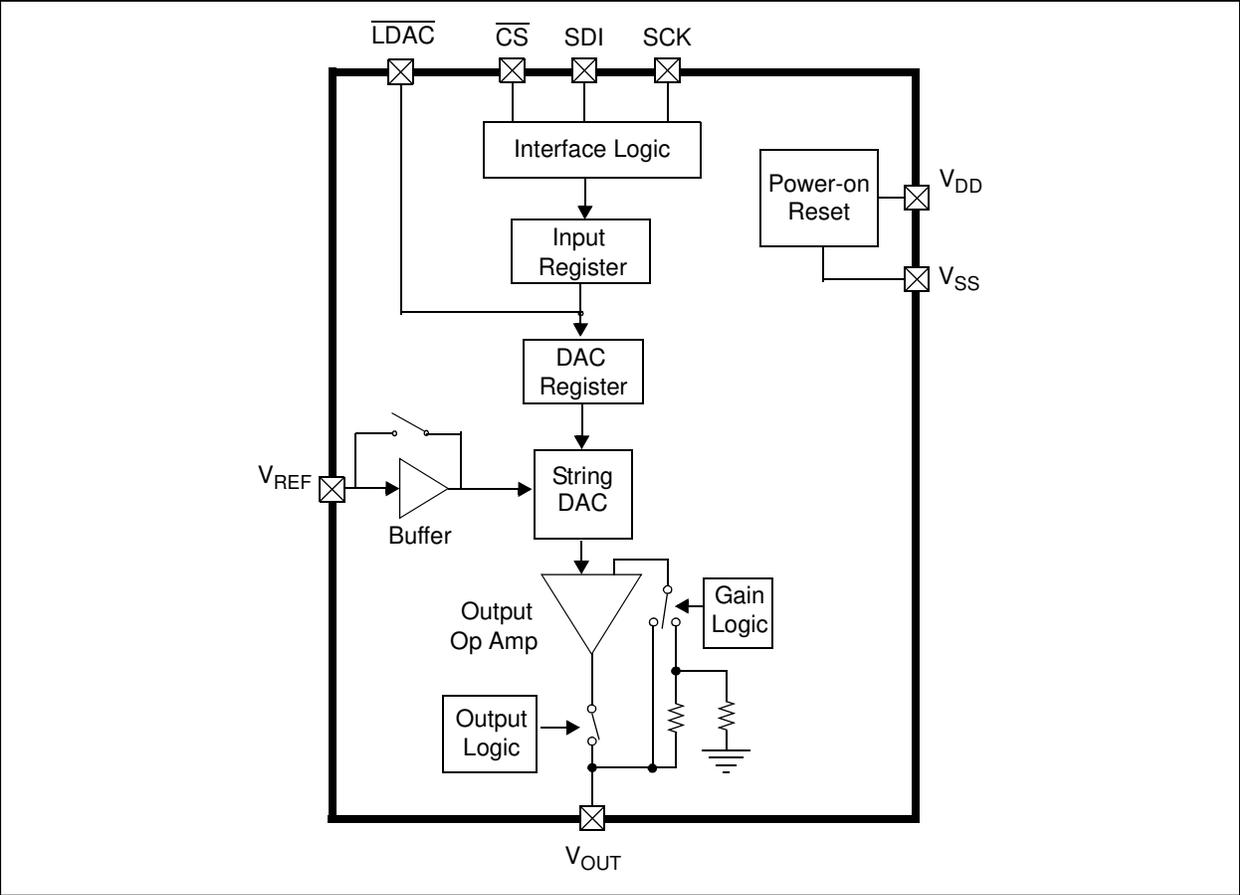
The MCP4901/4911/4921 devices are available in the PDIP, SOIC, MSOP and DFN packages.

Package Types



MCP4901/4911/4921

Block Diagram



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

V_{DD}	6.5V
All inputs and outputs w.r.t V_{SS}	-0.3V to $V_{DD}+0.3V$
Current at Input Pins	± 2 mA
Current at Supply Pins	± 50 mA
Current at Output Pins	± 25 mA
Storage temperature	-65°C to +150°C
Ambient temp. with power applied	-55°C to +125°C
ESD protection on all pins	≥ 4 kV (HBM), $\geq 400V$ (MM)
Maximum Junction Temperature (T_J)	+150°C

† **Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{REF} = 2.048V$, Output Buffer Gain (G) = 2x, $R_L = 5$ k Ω to GND, $C_L = 100$ pF $T_A = -40$ to +85°C. Typical values are at +25°C.

Parameters	Sym	Min	Typ	Max	Units	Conditions
Power Requirements						
Operating Voltage	V_{DD}	2.7	—	5.5		
Supply Current	I_{DD}	—	175	350	μA	$V_{DD} = 5V$ $V_{DD} = 3V$ V_{REF} input is unbuffered, all digital inputs are grounded, all analog outputs (V_{OUT}) are unloaded. Code = 0x000h
		—	125	250	μA	
Software Shutdown Current	I_{SHDN_SW}	—	3.3	6	μA	Power-on Reset circuit remains on
Power-On-Reset Threshold	V_{POR}	—	2.0	—	V	
DC Accuracy						
MCP4901						
Resolution	n	8	—	—	Bits	
INL Error	INL	-1	± 0.125	1	LSb	
DNL	DNL	-0.5	± 0.1	+0.5	LSb	Note 1
MCP4911						
Resolution	n	10	—	—	Bits	
INL Error	INL	-3.5	± 0.5	3.5	LSb	
DNL	DNL	-0.5	± 0.1	+0.5	LSb	Note 1
MCP4921						
Resolution	n	12	—	—	Bits	
INL Error	INL	-12	± 2	12	LSb	
DNL	DNL	-0.75	± 0.2	+0.75	LSb	Note 1

Note 1: Guaranteed monotonic by design over all codes.

2: This parameter is ensured by design, and not 100% tested.

MCP4901/4911/4921

ELECTRICAL CHARACTERISTICS (CONTINUED)

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{REF} = 2.048V$, Output Buffer Gain (G) = 2x, $R_L = 5\text{ k}\Omega$ to GND, $C_L = 100\text{ pF}$ $T_A = -40$ to $+85^\circ\text{C}$. Typical values are at $+25^\circ\text{C}$.						
Parameters	Sym	Min	Typ	Max	Units	Conditions
Offset Error	V_{OS}	—	± 0.02	1	% of FSR	Code = 0x000h
Offset Error Temperature Coefficient	$V_{OS}/^\circ\text{C}$	—	0.16	—	ppm/ $^\circ\text{C}$	-45°C to 25°C
		—	-0.44	—	ppm/ $^\circ\text{C}$	$+25^\circ\text{C}$ to 85°C
Gain Error	g_E	—	-0.10	1	% of FSR	Code = 0xFFh, not including offset error
Gain Error Temperature Coefficient	$\Delta G/^\circ\text{C}$	—	-3	—	ppm/ $^\circ\text{C}$	
Input Amplifier (V_{REF} Input)						
Input Range – Buffered Mode	V_{REF}	0.040	—	$V_{DD} - 0.040$	V	Note 2 Code = 2048
Input Range – Unbuffered Mode	V_{REF}	0	—	V_{DD}	V	$V_{REF} = 0.2\text{ Vp-p}$, $f = 100\text{ Hz}$ and 1 kHz
Input Impedance	R_{VREF}	—	165	—	$\text{k}\Omega$	Unbuffered Mode
Input Capacitance – Unbuffered Mode	C_{VREF}	—	7	—	pF	
Multiplier Mode -3 dB Bandwidth	f_{VREF}	—	450	—	kHz	$V_{REF} = 2.5V \pm 0.2V_{p-p}$, Unbuffered, $G = 1$
	f_{VREF}	—	400	—	kHz	$V_{REF} = 2.5V \pm 0.2V_{p-p}$, Unbuffered, $G = 2$
Multiplier Mode – Total Harmonic Distortion	THD_{VREF}	—	-73	—	dB	$V_{REF} = 2.5V \pm 0.2V_{p-p}$, Frequency = 1 kHz
Output Amplifier						
Output Swing	V_{OUT}	—	0.01 to $V_{DD} - 0.04$	—	V	Accuracy is better than 1 LSB for $V_{OUT} = 10\text{ mV}$ to $(V_{DD} - 40\text{ mV})$
Phase Margin	θ_m	—	66	—	Degrees	
Slew Rate	SR	—	0.55	—	V/ μs	
Short Circuit Current	I_{SC}	—	15	24	mA	
Settling Time	t_{settling}	—	4.5	—	μs	Within 1/2 LSB of final value from 1/4 to 3/4 full-scale range
Dynamic Performance (Note 2)						
DAC-to-DAC Crosstalk		—	10	—	nV-s	
Major Code Transition Glitch		—	45	—	nV-s	1 LSB change around major carry (0111...1111 to 1000...0000)
Digital Feedthrough		—	10	—	nV-s	
Analog Crosstalk		—	10	—	nV-s	

Note 1: Guaranteed monotonic by design over all codes.

2: This parameter is ensured by design, and not 100% tested.

ELECTRICAL CHARACTERISTIC WITH EXTENDED TEMPERATURE

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{REF} = 2.048V$, Output Buffer Gain (G) = 2x, $R_L = 5\text{ k}\Omega$ to GND, $C_L = 100\text{ pF}$. Typical values are at +125°C by characterization or simulation.

Parameters	Sym	Min	Typ	Max	Units	Conditions
Power Requirements						
Input Voltage	V_{DD}	2.7	—	5.5		
Input Current	I_{DD}	—	200	—	μA	V_{REF} input is unbuffered, all digital inputs are grounded, all analog outputs (VOUT) are unloaded. Code = 0x000h
Software Shutdown Current	I_{SHDN_SW}	—	5	—	μA	
Power-on Reset Threshold	V_{POR}	—	1.85	—	V	
DC Accuracy						
MCP4901						
Resolution	n	8	—	—	Bits	
INL Error	INL		± 0.25		LSb	
DNL	DNL		± 0.2		LSb	Note 1
MCP4911						
Resolution	n	10	—	—	Bits	
INL Error	INL		± 1		LSb	
DNL	DNL		± 0.2		LSb	Note 1
MCP4921						
Resolution	n	12	—	—	Bits	
INL Error	INL		± 4		LSb	
DNL	DNL		± 0.25		LSb	Note 1
Offset Error	V_{OS}	—	± 0.02	—	% of FSR	Code = 0x000h
Offset Error Temperature Coefficient	$V_{OS}/^{\circ}C$	—	-5	—	ppm/ $^{\circ}C$	+25°C to +125°C
Gain Error	g_E	—	-0.10	—	% of FSR	Code = 0xFFFFh, not including offset error
Gain Error Temperature Coefficient	$\Delta G/^{\circ}C$	—	-3	—	ppm/ $^{\circ}C$	
Input Amplifier (V_{REF} Input)						
Input Range – Buffered Mode	V_{REF}	—	0.040 to $V_{DD} - 0.040$	—	V	Note 1 Code = 2048, $V_{REF} = 0.2\text{ Vp-p}$, $f = 100\text{ Hz}$ and 1 kHz
Input Range – Unbuffered Mode	V_{REF}	0	—	V_{DD}	V	
Input Impedance	R_{VREF}	—	174	—	k Ω	Unbuffered Mode
Input Capacitance – Unbuffered Mode	C_{VREF}	—	7	—	pF	
Multiplying Mode -3 dB Bandwidth	f_{VREF}	—	450	—	kHz	$V_{REF} = 2.5V \pm 0.1\text{ Vp-p}$, Unbuffered, G = 1x
	f_{VREF}	—	400	—	kHz	$V_{REF} = 2.5V \pm 0.1\text{ Vp-p}$, Unbuffered, G = 2x

Note 1: Guaranteed monotonic by design over all codes.

2: This parameter is ensured by design, and not 100% tested.

MCP4901/4911/4921

ELECTRICAL CHARACTERISTIC WITH EXTENDED TEMPERATURE (CONTINUED)

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{REF} = 2.048V$, Output Buffer Gain (G) = 2x, $R_L = 5\text{ k}\Omega$ to GND, $C_L = 100\text{ pF}$. Typical values are at +125°C by characterization or simulation.						
Parameters	Sym	Min	Typ	Max	Units	Conditions
Multiplying Mode - Total Harmonic Distortion	$THD_{V_{REF}}$	—	—	—	dB	$V_{REF} = 2.5V \pm 0.1V_{p-p}$, Frequency = 1 kHz
Output Amplifier						
Output Swing	V_{OUT}	—	0.01 to $V_{DD} - 0.04$	—	V	Accuracy is better than 1 LSB for $V_{OUT} = 10\text{ mV}$ to $(V_{DD} - 40\text{ mV})$
Phase Margin	θ_m	—	66	—	Degrees	
Slew Rate	SR	—	0.55	—	V/ μ s	
Short Circuit Current	I_{SC}	—	17	—	mA	
Settling Time	$t_{settling}$	—	4.5	—	μ s	Within 1/2 LSB of final value from 1/4 to 3/4 full-scale range
Dynamic Performance (Note 2)						
Major Code Transition Glitch		—	45	—	nV-s	1 LSB change around major carry (0111...1111 to 1000...0000)
Digital Feedthrough		—	10	—	nV-s	

Note 1: Guaranteed monotonic by design over all codes.

2: This parameter is ensured by design, and not 100% tested.

AC CHARACTERISTICS (SPI TIMING SPECIFICATIONS)

Electrical Specifications: Unless otherwise indicated, $V_{DD} = 2.7V - 5.5V$, $T_A = -40$ to $+125^\circ C$. Typical values are at $+25^\circ C$.

Parameters	Sym	Min	Typ	Max	Units	Conditions
Schmitt Trigger High Level Input Voltage (All digital input pins)	V_{IH}	$0.7 V_{DD}$	—	—	V	
Schmitt Trigger Low Level Input Voltage (All digital input pins)	V_{IL}	—	—	$0.2 V_{DD}$	V	
Hysteresis of Schmitt Trigger Inputs	V_{HYS}	—	$0.05 V_{DD}$	—		
Input Leakage Current	$I_{LEAKAGE}$	-1	—	1	μA	$\overline{LDAC} = \overline{CS} = \overline{SDI} = \overline{SCK} = V_{REF} = V_{DD}$ or V_{SS}
Digital Pin Capacitance (All inputs/outputs)	C_{IN}, C_{OUT}	—	10	—	pF	$V_{DD} = 5.0V$, $T_A = +25^\circ C$, $f_{CLK} = 1$ MHz (Note 1)
Clock Frequency	F_{CLK}	—	—	20	MHz	$T_A = +25^\circ C$ (Note 1)
Clock High Time	t_{HI}	15	—	—	ns	Note 1
Clock Low Time	t_{LO}	15	—	—	ns	Note 1
\overline{CS} Fall to First Rising CLK Edge	t_{CSSR}	40	—	—	ns	Applies only when \overline{CS} falls with CLK high (Note 1)
Data Input Setup Time	t_{SU}	15	—	—	ns	Note 1
Data Input Hold Time	t_{HD}	10	—	—	ns	Note 1
SCK Rise to \overline{CS} Rise Hold Time	t_{CHS}	15	—	—	ns	Note 1
\overline{CS} High Time	t_{CSH}	15	—	—	ns	Note 1
\overline{LDAC} Pulse Width	t_{LD}	100	—	—	ns	Note 1
\overline{LDAC} Setup Time	t_{LS}	40	—	—	ns	Note 1
SCK Idle Time before \overline{CS} Fall	t_{IDLE}	40	—	—	ns	Note 1

Note 1: This parameter is ensured by design and not 100% tested.

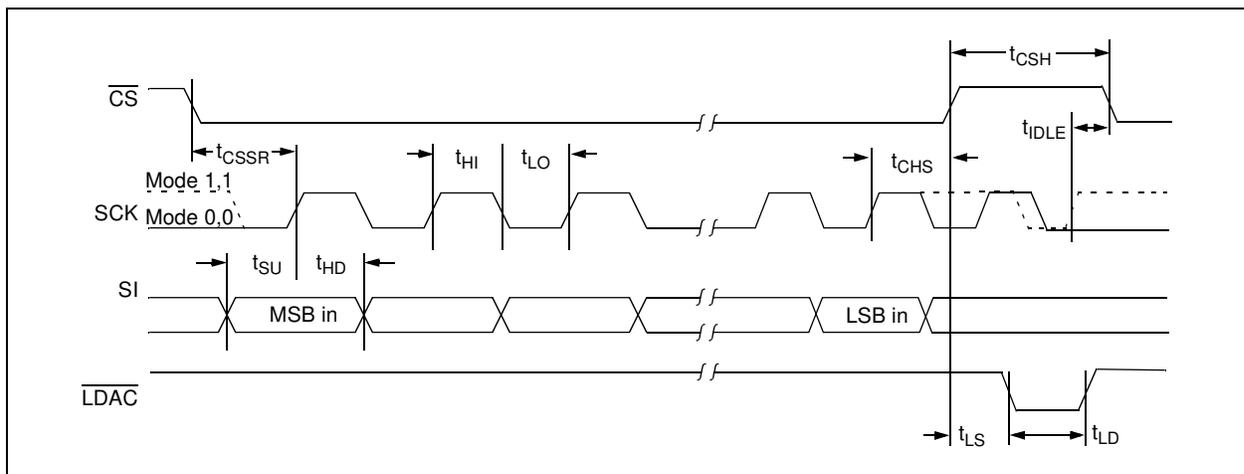


FIGURE 1-1: SPI Input Timing Data.

MCP4901/4911/4921

TEMPERATURE CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, $V_{DD} = +2.7V$ to $+5.5V$, $V_{SS} = GND$.						
Parameters	Sym	Min	Typ	Max	Units	Conditions
Temperature Ranges						
Specified Temperature Range	T_A	-40	—	+125	°C	
Operating Temperature Range	T_A	-40	—	+125	°C	Note 1
Storage Temperature Range	T_A	-65	—	+150	°C	
Thermal Package Resistances						
Thermal Resistance, 8L-DFN (2 x 3)	θ_{JA}	—	68	—	°C/W	
Thermal Resistance, 8L-PDIP	θ_{JA}	—	90	—	°C/W	
Thermal Resistance, 8L-SOIC	θ_{JA}	—	150	—	°C/W	
Thermal Resistance, 8L-MSOP	θ_{JA}	—	211	—	°C/W	

Note 1: The MCP4901/4911/4921 devices operate over this extended temperature range, but with reduced performance. Operation in this range must not cause T_J to exceed the maximum junction temperature of 150°C.

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, $\text{Gain} = 2\text{x}$, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

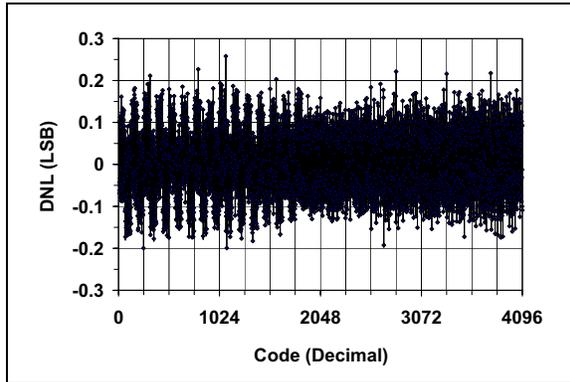


FIGURE 2-1: DNL vs. Code (MCP4921).

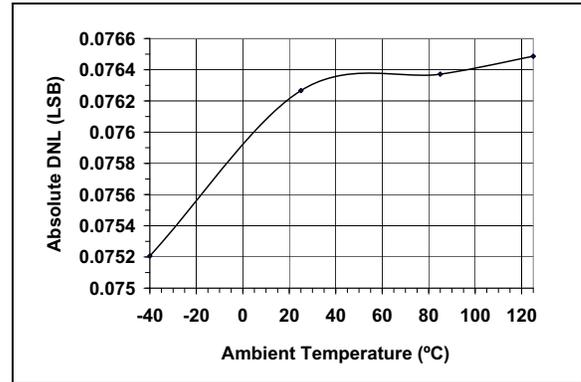


FIGURE 2-4: Absolute DNL vs. Temperature (MCP4921).

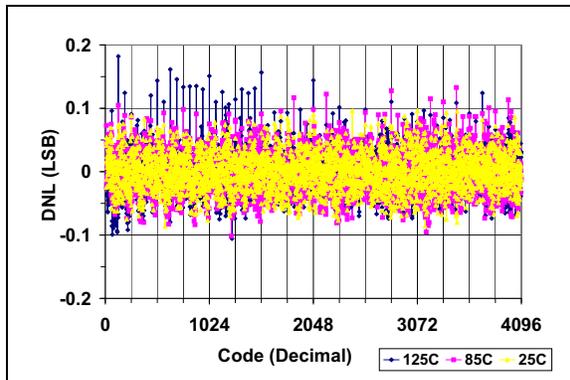


FIGURE 2-2: DNL vs. Code and Temperature (MCP4921).

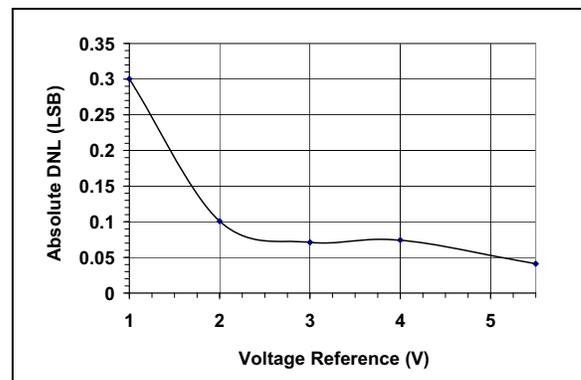


FIGURE 2-5: Absolute DNL vs. Voltage Reference (MCP4921).

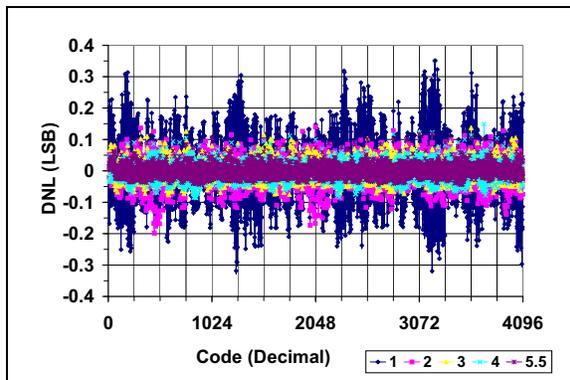


FIGURE 2-3: DNL vs. Code and V_{REF} Gain=1 (MCP4921).

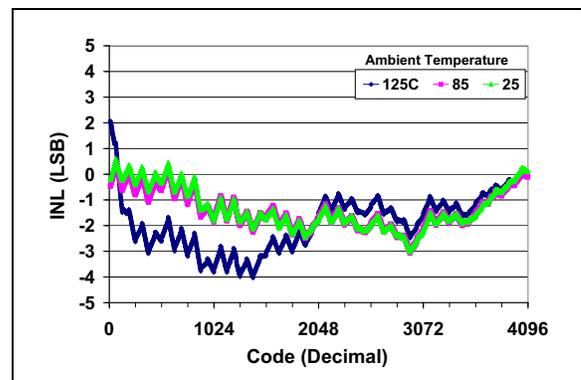


FIGURE 2-6: INL vs. Code and Temperature (MCP4921).

MCP4901/4911/4921

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

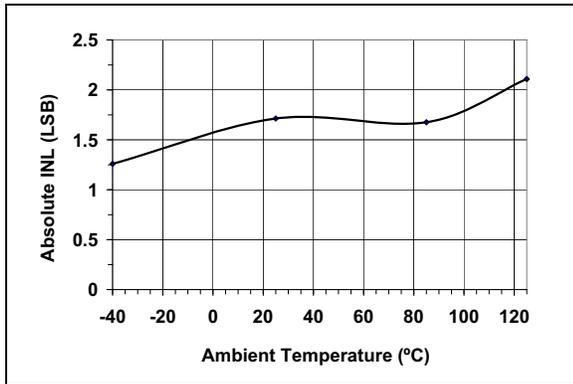


FIGURE 2-7: Absolute INL vs. Temperature (MCP4921).

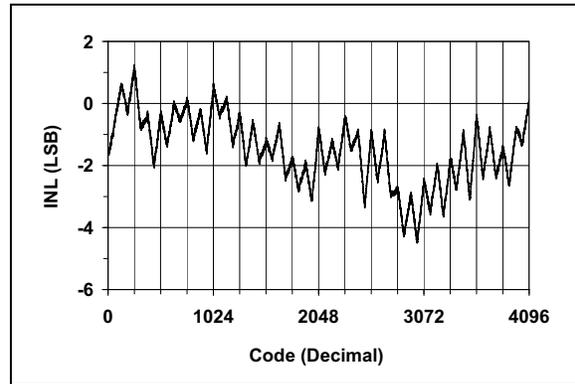


FIGURE 2-10: INL vs. Code (MCP4921).

Note: Single device graph (Figure 2-10) for illustration of 64 code effect.

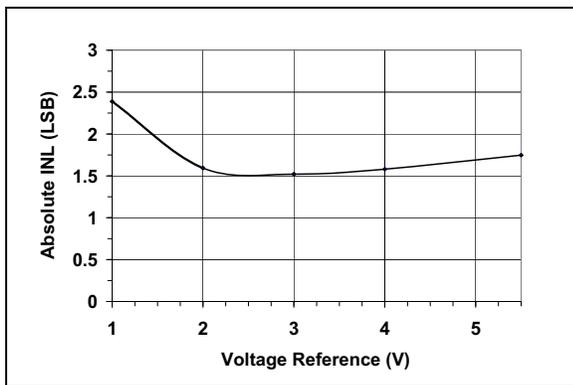


FIGURE 2-8: Absolute INL vs. V_{REF} (MCP4921).

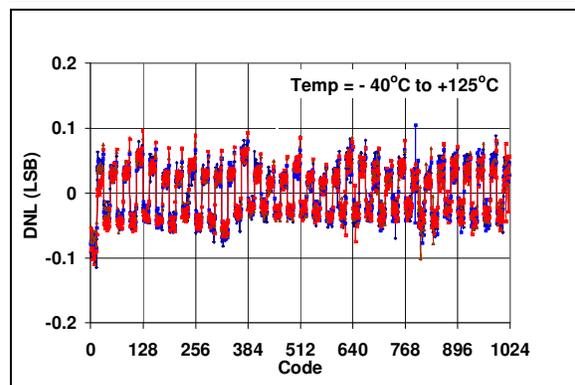


FIGURE 2-11: DNL vs. Code and Temperature (MCP4911).

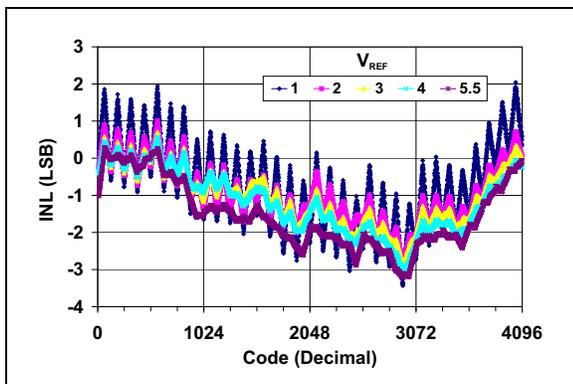


FIGURE 2-9: INL vs. Code and V_{REF} (MCP4921).

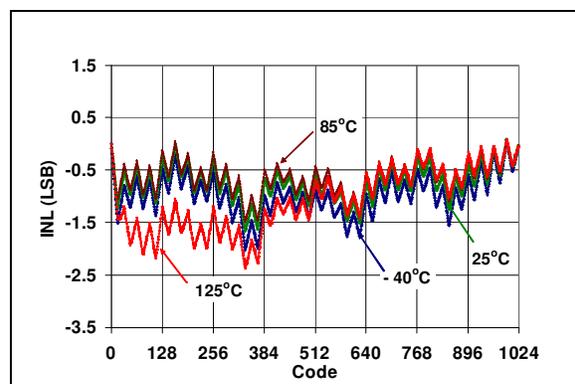


FIGURE 2-12: INL vs. Code and Temperature (MCP4911).

MCP4901/4911/4921

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

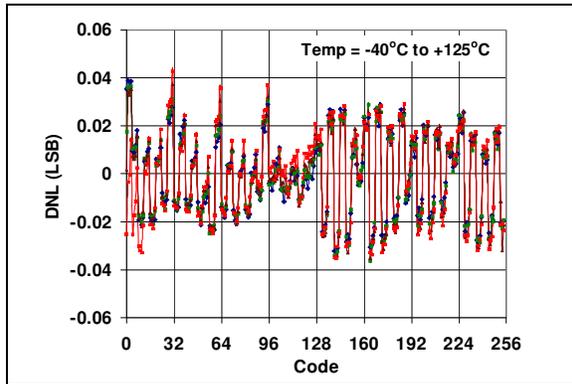


FIGURE 2-13: DNL vs. Code and Temperature (MCP4901).

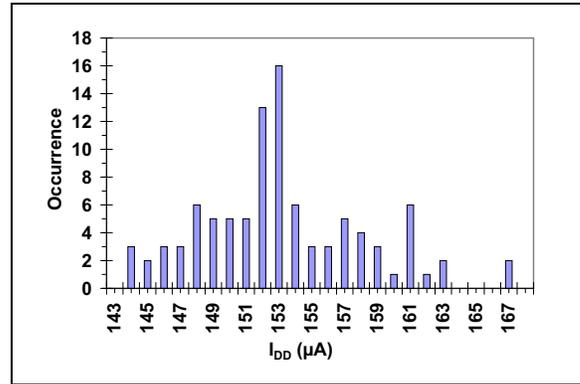


FIGURE 2-16: I_{DD} Histogram ($V_{DD} = 2.7\text{V}$).

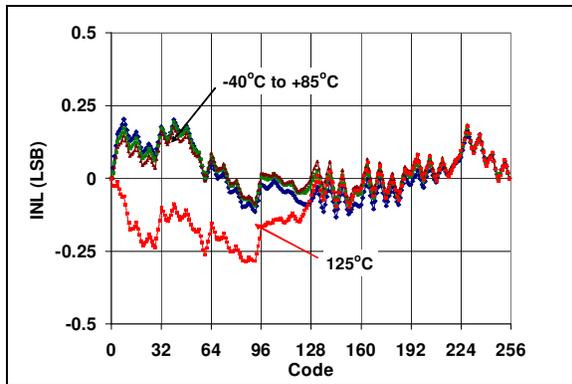


FIGURE 2-14: INL vs. Code and Temperature (MCP4901).

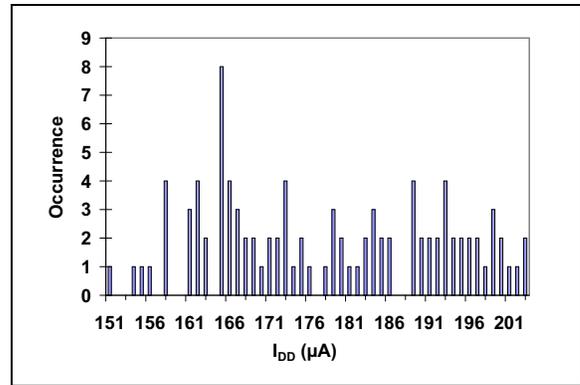


FIGURE 2-17: I_{DD} Histogram ($V_{DD} = 5.0\text{V}$).

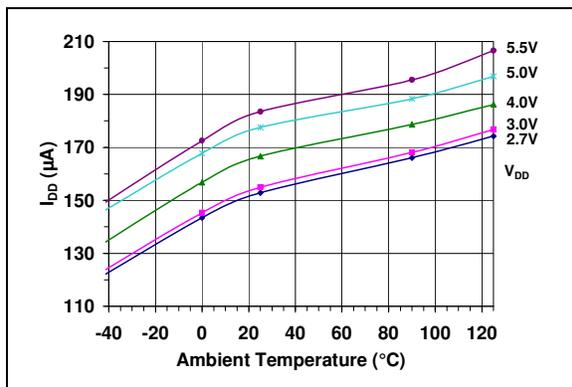


FIGURE 2-15: I_{DD} vs. Temperature and V_{DD} .

MCP4901/4911/4921

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

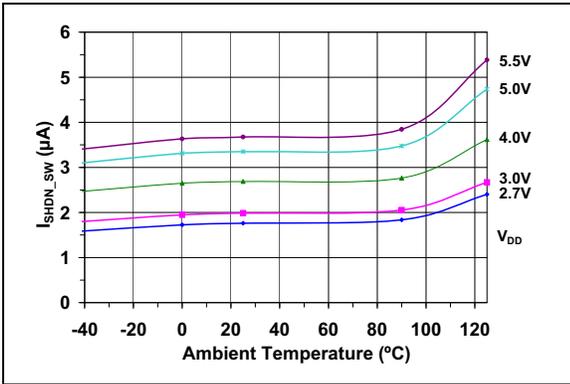


FIGURE 2-18: Shutdown Current vs. Temperature and V_{DD} .

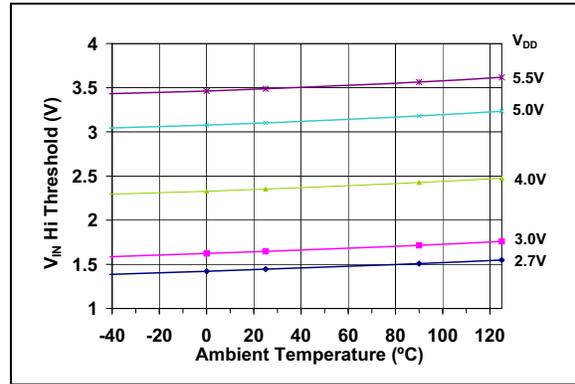


FIGURE 2-21: V_{IN} High Threshold vs. Temperature and V_{DD} .

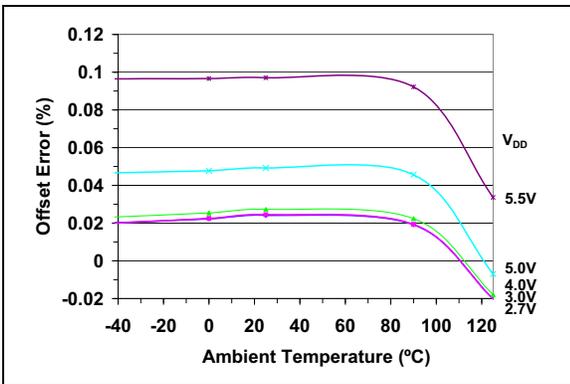


FIGURE 2-19: Offset Error vs. Temperature and V_{DD} .

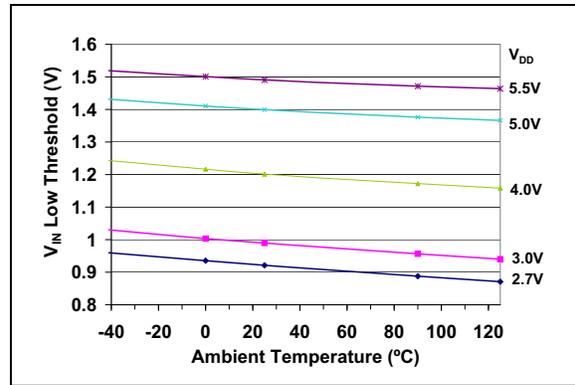


FIGURE 2-22: V_{IN} Low Threshold vs. Temperature and V_{DD} .

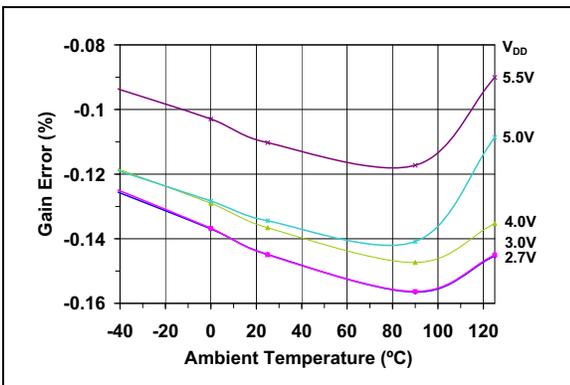


FIGURE 2-20: Gain Error vs. Temperature and V_{DD} .

MCP4901/4911/4921

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

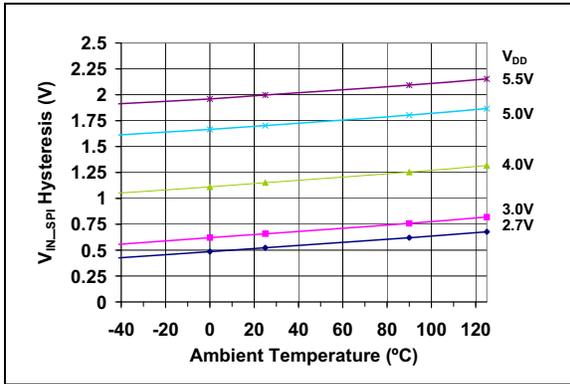


FIGURE 2-23: Input Hysteresis vs. Temperature and V_{DD} .

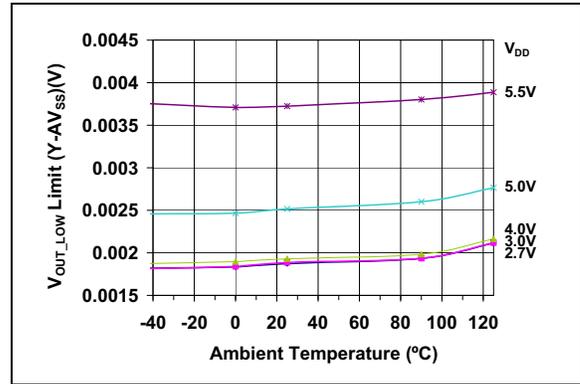


FIGURE 2-26: V_{OUT} Low Limit vs. Temperature and V_{DD} .

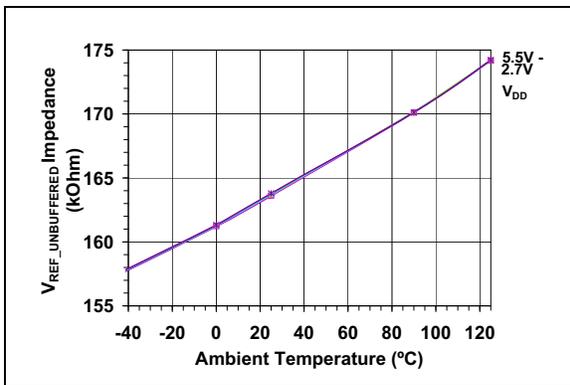


FIGURE 2-24: V_{REF} Input Impedance vs. Temperature and V_{DD} .

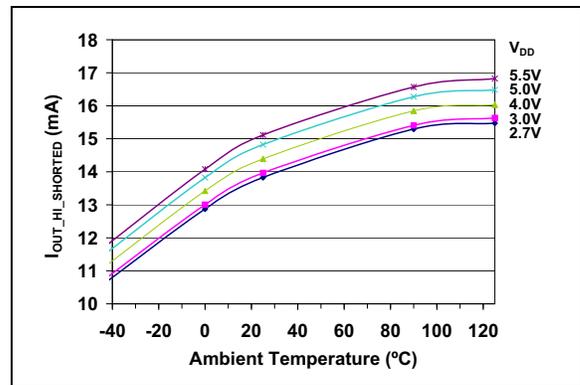


FIGURE 2-27: I_{OUT} High Short vs. Temperature and V_{DD} .

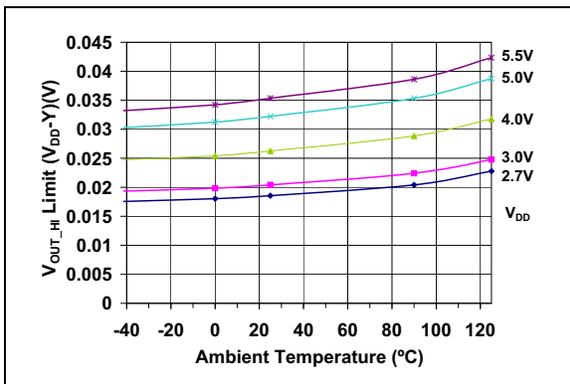


FIGURE 2-25: V_{OUT} High Limit vs. Temperature and V_{DD} .

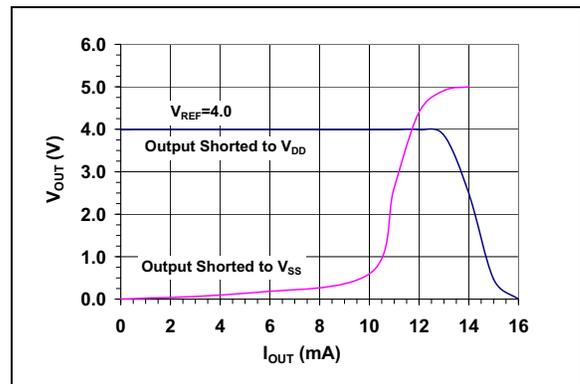


FIGURE 2-28: I_{OUT} vs. V_{OUT} . Gain = 1.

MCP4901/4911/4921

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.048\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

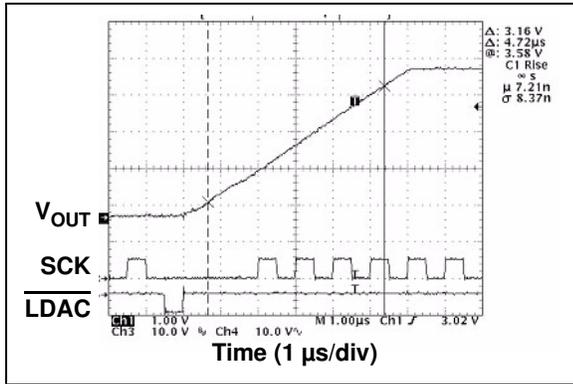


FIGURE 2-29: V_{OUT} Rise Time

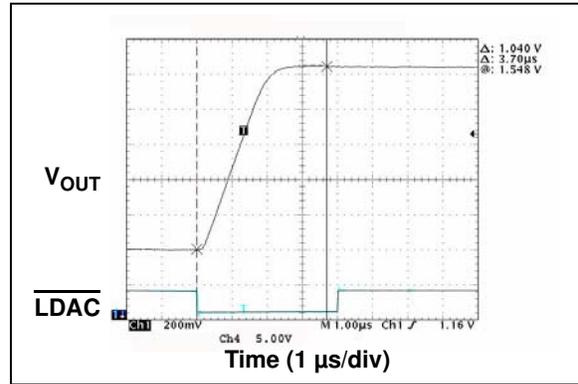


FIGURE 2-32: V_{OUT} Rise Time

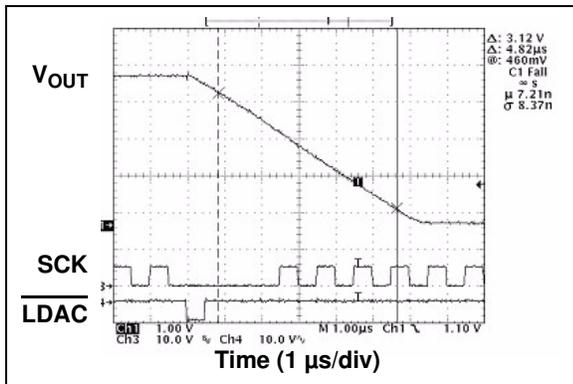


FIGURE 2-30: V_{OUT} Fall Time.

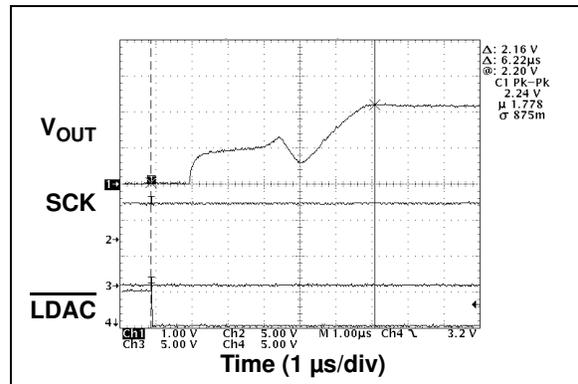


FIGURE 2-33: V_{OUT} Rise Time Exit Shutdown.

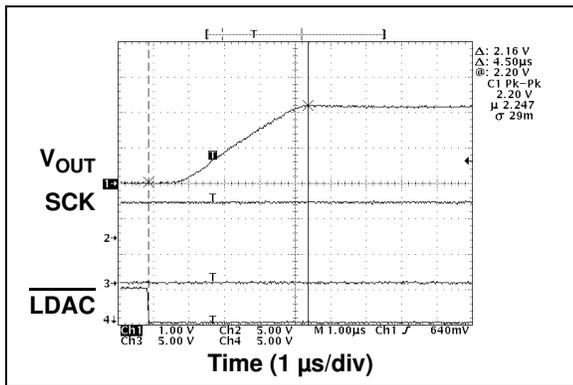


FIGURE 2-31: V_{OUT} Rise Time

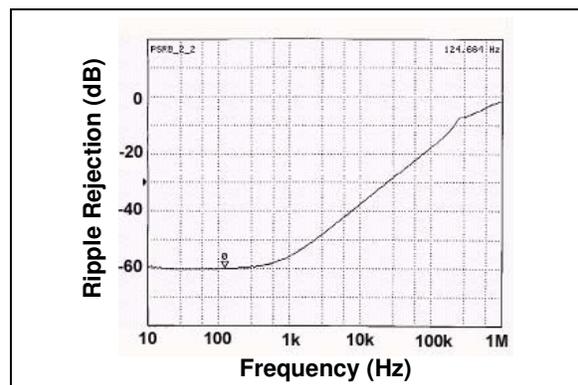


FIGURE 2-34: PSRR vs. Frequency.

Note: Unless otherwise indicated, $T_A = +25^\circ\text{C}$, $V_{DD} = 5\text{V}$, $V_{SS} = 0\text{V}$, $V_{REF} = 2.50\text{V}$, Gain = 2, $R_L = 5\text{ k}\Omega$, $C_L = 100\text{ pF}$.

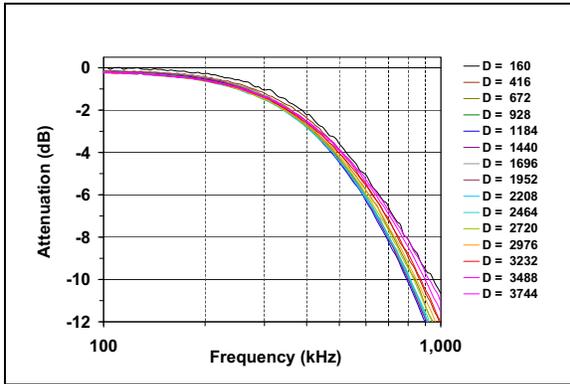


FIGURE 2-35: Multiplier Mode Bandwidth.

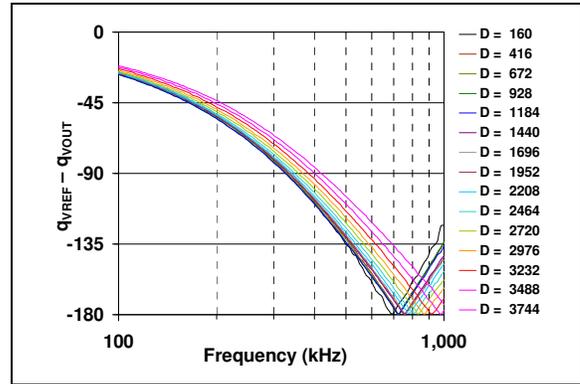


FIGURE 2-37: Phase Shift.

Figure 2-35 calculation:

$$\text{Attenuation (dB)} = 20 \log (V_{\text{OUT}}/V_{\text{REF}}) - 20 \log (G(D/4096))$$

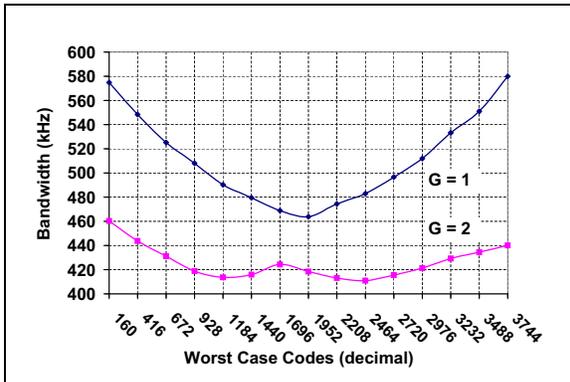


FIGURE 2-36: -3 db Bandwidth vs. Worst Codes.

MCP4901/4911/4921

NOTES:

3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in [Table 3-1](#).

TABLE 3-1: PIN FUNCTION TABLE

PDIP, MSOP, SOIC	DFN	Symbol	Description
1	1	V_{DD}	Supply Voltage Input (2.7V to 5.5V)
2	2	\overline{CS}	Chip Select Input
3	3	SCK	Serial Clock Input
4	4	SDI	Serial Data Input
5	5	\overline{LDAC}	DAC Output Synchronization Input. This pin is used to transfer the input register (DAC settings) to the output register (V_{OUT})
6	6	V_{REF}	Voltage Reference Input
7	7	V_{SS}	Ground reference point for all circuitry on the device
8	8	V_{OUT}	DAC Analog Output
—	9	EP	Exposed Thermal Pad. This pad must be connected to V_{SS} in application

3.1 Supply Voltage Pins (V_{DD} , V_{SS})

V_{DD} is the positive supply voltage input pin. The input supply voltage is relative to V_{SS} and can range from 2.7V to 5.5V. The power supply at the V_{DD} pin should be as clean as possible for good DAC performance. It is recommended to use an appropriate bypass capacitor of about 0.1 μ F (ceramic) to ground. An additional 10 μ F capacitor (tantalum) in parallel is also recommended to further attenuate high-frequency noise present in application boards.

V_{SS} is the analog ground pin and the current return path of the device. The user must connect the V_{SS} pin to a ground plane through a low-impedance connection. If an analog ground path is available in the application Printed Circuit Board (PCB), it is highly recommended that the V_{SS} pin be tied to the analog ground path or isolated within an analog ground plane of the circuit board.

3.2 Chip Select (\overline{CS})

\overline{CS} is the chip select input, which requires an active-low signal to enable serial clock and data functions.

3.3 Serial Clock Input (SCK)

SCK is the SPI compatible serial clock input.

3.4 Serial Data Input (SDI)

SDI is the SPI compatible serial data input.

3.5 Latch DAC Input (\overline{LDAC})

The \overline{LDAC} (latch DAC synchronization input) pin is used to transfer the input latch register to the DAC register (output latches, V_{OUT}). When this pin is low, V_{OUT} is updated with input register content. This pin can be tied to low (V_{SS}) if the V_{OUT} update is desired at the rising edge of the \overline{CS} pin. This pin can be driven by an external control device such as an MCU I/O pin.

3.6 Analog Output (V_{OUT})

V_{OUT} is the DAC analog output pin. The DAC output has an output amplifier. The full-scale range of the DAC output is from V_{SS} to $G \cdot V_{REF}$, where G is the gain selection option (1x or 2x). The DAC analog output cannot go higher than the supply voltage (V_{DD}).

3.7 Voltage Reference Input (V_{REF})

V_{REF} is the voltage reference input for the device. The reference on this pin is utilized to set the reference voltage on the string DAC. The input voltage can range from V_{SS} to V_{DD} . This pin can be tied to V_{DD} .

3.8 Exposed Thermal Pad (EP)

There is an internal electrical connection between the Exposed Thermal Pad (EP) and the V_{SS} pin. They must be connected to the same potential on the PCB.

MCP4901/4911/4921

NOTES:

4.0 GENERAL OVERVIEW

The MCP4901, MCP4911 and MCP4921 are single channel voltage output 8-bit, 10-bit and 12-bit DAC devices, respectively. These devices include a V_{REF} input buffer, a rail-to-rail output amplifier, shutdown and reset management circuitry. The devices use an SPI serial communication interface and operate with a single-supply voltage from 2.7V to 5.5V.

The DAC input coding of these devices is straight binary. Equation 4-1 shows the DAC analog output voltage calculation.

EQUATION 4-1: ANALOG OUTPUT VOLTAGE (V_{OUT})

$$V_{OUT} = \frac{(V_{REF} \times D_n)}{2^n} G$$

Where:

- V_{REF} = External voltage reference
- D_n = DAC input code
- G = Gain Selection
 - = 2 for $\langle \overline{GA} \rangle$ bit = 0
 - = 1 for $\langle \overline{GA} \rangle$ bit = 1
- n = DAC Resolution
 - = 8 for MCP4901
 - = 10 for MCP4911
 - = 12 for MCP4912

The ideal output range of each device is:

- **MCP4901 (n = 8)**
 - (a) 0V to $255/256 \times V_{REF}$ when gain setting = 1x.
 - (b) 0V to $255/256 \times 2 \times V_{REF}$ when gain setting = 2x.
- **MCP4911 (n = 10)**
 - (a) 0V to $1023/1024 \times V_{REF}$ when gain setting = 1x.
 - (b) 0V to $1023/1024 \times 2 \times V_{REF}$ when gain setting = 2x.
- **MCP4921 (n = 12)**
 - (a) 0V to $4095/4096 \times V_{REF}$ when gain setting = 1x.
 - (b) 0V to $4095/4096 \times 2 \times V_{REF}$ when gain setting = 2x.

Note: See the output swing voltage specification in Section 1.0 “Electrical Characteristics”.

1 LSB is the ideal voltage difference between two successive codes. Table 4-1 illustrates the LSB calculation of each device.

TABLE 4-1: LSb OF EACH DEVICE

Device	Gain Selection	LSb Size
MCP4901 (n = 8)	1x	$V_{REF}/256$
	2x	$(2 \times V_{REF})/256$
MCP4911 (n = 10)	1x	$V_{REF}/1024$
	2x	$(2 \times V_{REF})/1024$
MCP4921 (n = 12)	1x	$V_{REF}/4096$
	2x	$(2 \times V_{REF})/4096$

where V_{REF} is the external voltage reference.

4.1 DC Accuracy

4.1.1 INL ACCURACY

Integral Non-Linearity (INL) error is the maximum deviation between an actual code transition point and its corresponding ideal transition point, after offset and gain errors have been removed. The two endpoints (from 0x000 and 0xFFFF) method is used for the calculation. Figure 4-1 shows the details.

A positive INL error represents transition(s) later than ideal. A negative INL error represents transition(s) earlier than ideal.

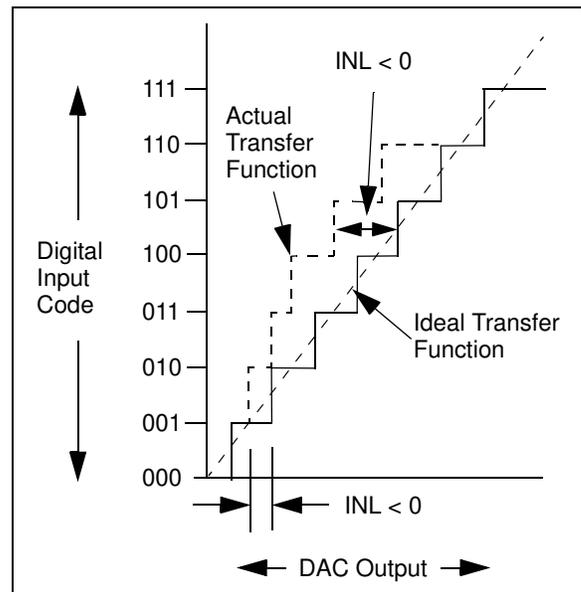


FIGURE 4-1: Example for INL Error.

4.1.2 DNL ACCURACY

A Differential Non-Linearity (DNL) error is the measure of variations in code widths from the ideal code width. A DNL error of zero indicates that every code is exactly 1 LSB wide.

MCP4901/4911/4921

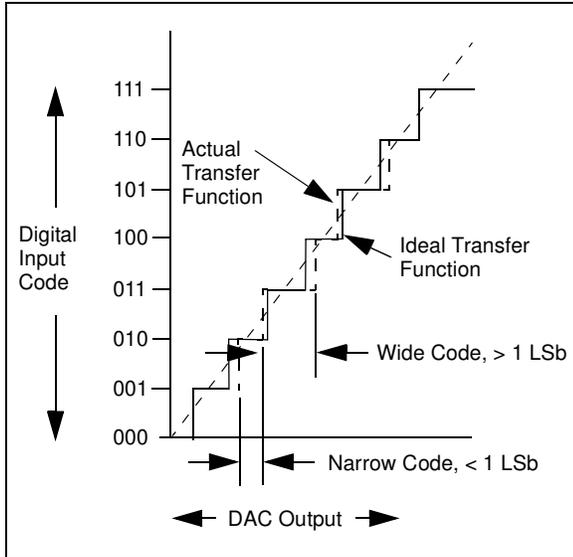


FIGURE 4-2: Example for DNL Accuracy.

4.1.3 OFFSET ERROR

An offset error is the deviation from zero voltage output when the digital input code is zero.

4.1.4 GAIN ERROR

A gain error is the deviation from the ideal output, $V_{REF} - 1 \text{ LSB}$, excluding the effects of offset error.

4.2 Circuit Descriptions

4.2.1 OUTPUT AMPLIFIER

The DAC's output is buffered with a low-power, precision CMOS amplifier. This amplifier provides low offset voltage and low noise. The output stage enables the device to operate with output voltages close to the power supply rails. Refer to **Section 1.0 "Electrical Characteristics"** for the analog output voltage range and load conditions.

In addition to resistive load driving capability, the amplifier will also drive high capacitive loads without oscillation. The amplifier's strong output allows V_{OUT} to be used as a programmable voltage reference in a system.

Selecting a gain of 2 reduces the bandwidth of the amplifier in Multiplying mode. Refer to **Section 1.0 "Electrical Characteristics"** for the Multiplying mode bandwidth for given load conditions.

4.2.1.1 Programmable Gain Block

The rail-to-rail output amplifier has two configurable gain options: a gain of 1x ($\langle \overline{GA} \rangle = 1$) or a gain of 2x ($\langle \overline{GA} \rangle = 0$). The default value is a gain of 2x ($\langle \overline{GA} \rangle = 0$).

4.2.2 VOLTAGE REFERENCE AMPLIFIER

The input buffer amplifier for the MCP4901/4911/4921 devices provides low offset voltage and low noise. A Configuration bit for each DAC allows the V_{REF} input to bypass the V_{REF} input buffer amplifier, achieving Buffered or Unbuffered mode. Buffered mode provides a very high input impedance, with only minor limitations on the input range and frequency response. Unbuffered mode provides a wide input range (0V to V_{DD}), with a typical input impedance of 165 k Ω with 7 pF. Unbuffered mode ($\langle \overline{BUF} \rangle = 0$) is the default configuration.

4.2.3 POWER-ON RESET CIRCUIT

The internal Power-on Reset (POR) circuit monitors the power supply voltage (V_{DD}) during device operation. The circuit also ensures that the device powers up with high output impedance ($\langle \overline{SHDN} \rangle = 0$, typically 500 k Ω). The devices will continue to have a high-impedance output until a valid write command is received, and the \overline{LDAC} pin meets the input low threshold.

If the power supply voltage is less than the POR threshold ($V_{POR} = 2.0\text{V}$, typical), the device will be held in its Reset state. It will remain in that state until $V_{DD} > V_{POR}$ and a subsequent write command is received.

Figure 4-3 shows a typical power supply transient pulse and the duration required to cause a reset to occur, as well as the relationship between the duration and trip voltage. A 0.1 μF decoupling capacitor, mounted as close as possible to the V_{DD} pin, can provide additional transient immunity.

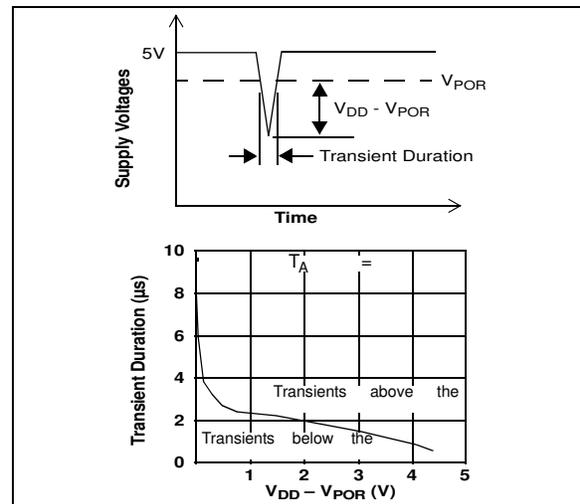


FIGURE 4-3: Typical Transient Response.

4.2.4 SHUTDOWN MODE

The user can shut down the device by using a software command. During Shutdown mode, most of the internal circuits, including the output amplifier, are turned off for power savings. The serial interface remains active, thus allowing a write command to bring the device out of Shutdown mode. There will be no analog output at the V_{OUT} pin, and the V_{OUT} pin is internally switched to a known resistive load (500 k Ω , typical). Figure 4-4 shows the analog output stage during Shutdown mode.

The device will remain in Shutdown mode until it receives a write command with $\langle \overline{SHDN} \rangle$ bit = 1 and the bit is latched into the device. When the device is changed from Shutdown to Active mode, the output settling time takes less than 10 μs , but more than the standard active mode settling time (4.5 μs).

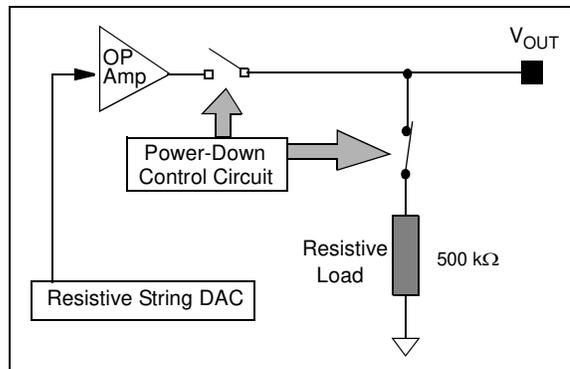


FIGURE 4-4: Output Stage for Shutdown Mode.

MCP4901/4911/4921

NOTES:

5.0 SERIAL INTERFACE

5.1 Overview

The MCP4901/4911/4921 devices are designed to interface directly with the Serial Peripheral Interface (SPI) port, which is available on many microcontrollers and supports Mode 0,0 and Mode 1,1. Commands and data are sent to the device via the SDI pin, with data being clocked-in on the rising edge of SCK. The communications are unidirectional, thus the data cannot be read out of the MCP4901/4911/4921. The \overline{CS} pin must be held low for the duration of a write command. The write command consists of 16 bits and is used to configure the DAC's control and data latches. Register 5-1 through Register 5-3 detail the input register that is used to configure and load the DAC register for each device. Figure 5-1 through Figure 5-3 show the write command for each device.

Refer to Figure 1-1 and the SPI Timing Specifications Table for detailed input and output timing specifications for both Mode 0,0 and Mode 1,1 operation.

5.2 Write Command

The write command is initiated by driving the \overline{CS} pin low, followed by clocking the four Configuration bits and the 12 data bits into the SDI pin on the rising edge of SCK. The \overline{CS} pin is then raised, causing the data to be latched into the DAC's input register.

The MCP4901/4911/4921 utilizes a double-buffered latch structure to allow the analog output to be synchronized with the \overline{LDAC} pin, if desired.

By bringing the \overline{LDAC} pin down to a low state, the content stored in the DAC's input register is transferred into the DAC's output register (V_{OUT}), and V_{OUT} is updated.

All writes to the MCP4901/4911/4921 devices are 16-bit words. Any clocks past the 16th clock will be ignored. The Most Significant 4 bits are Configuration bits. The remaining 12 bits are data bits. No data can be transferred into the device with \overline{CS} high. This transfer will only occur if 16 clocks have been transferred into the device. If the rising edge of \overline{CS} occurs prior to that, shifting of data into the input register will be aborted.

MCP4901/4911/4921

REGISTER 5-1: WRITE COMMAND REGISTER FOR MCP4921 (12-BIT DAC)

W-x	W-x	W-x	W-0	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x
0	BUF	\overline{GA}	\overline{SHDN}	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
bit 15								bit 0							

REGISTER 5-2: WRITE COMMAND REGISTER FOR MCP4911 (10-BIT DAC)

W-x	W-x	W-x	W-0	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x
0	BUF	\overline{GA}	\overline{SHDN}	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	x	x
bit 15								bit 0							

REGISTER 5-3: WRITE COMMAND REGISTER FOR MCP4901 (8-BIT DAC)

W-x	W-x	W-x	W-0	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x	W-x
0	BUF	\overline{GA}	\overline{SHDN}	D7	D6	D5	D4	D3	D2	D1	D0	x	x	x	x
bit 15								bit 0							

Where:

- bit 15 0 = Write to DAC register
 1 = Ignore this command
- bit 14 **BUF:** V_{REF} Input Buffer Control bit
 1 = Buffered
 0 = Unbuffered
- bit 13 **GA:** Output Gain Selection bit
 1 = $1x (V_{OUT} = V_{REF} * D/4096)$
 0 = $2x (V_{OUT} = 2 * V_{REF} * D/4096)$
- bit 12 **SHDN:** Output Shutdown Control bit
 1 = Active mode operation. V_{OUT} is available.
 0 = Shutdown the device. Analog output is not available. V_{OUT} pin is connected to 500 k Ω (typical).
- bit 11-0 **D11:D0:** DAC Input Data bits. Bit x is ignored.

Legend

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
-n = Value at POR	1 = bit is set	0 = bit is cleared x = bit is unknown

MCP4901/4911/4921

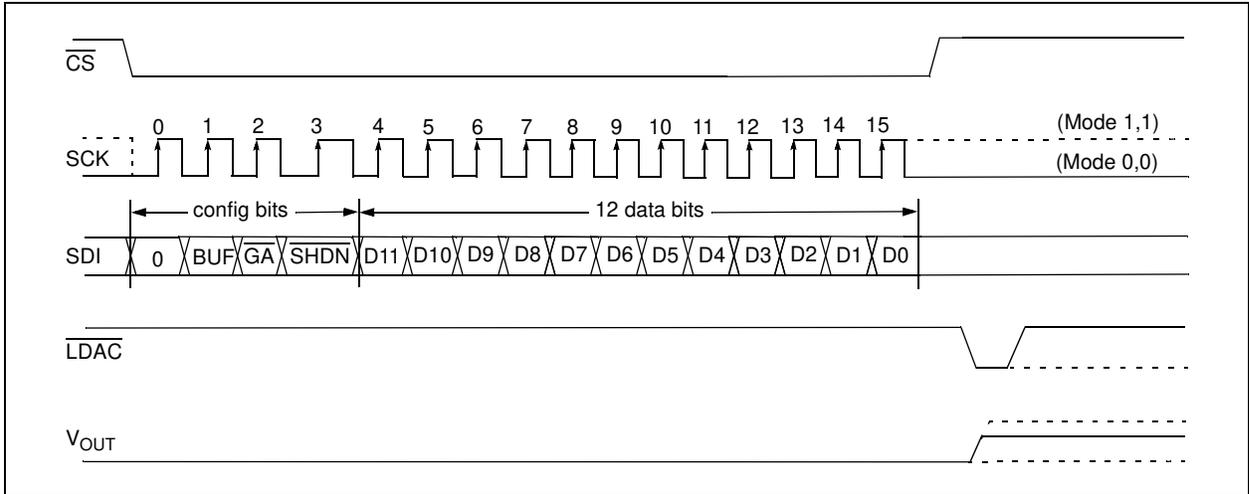


FIGURE 5-1: Write Command for MCP4921 (12-bit DAC).

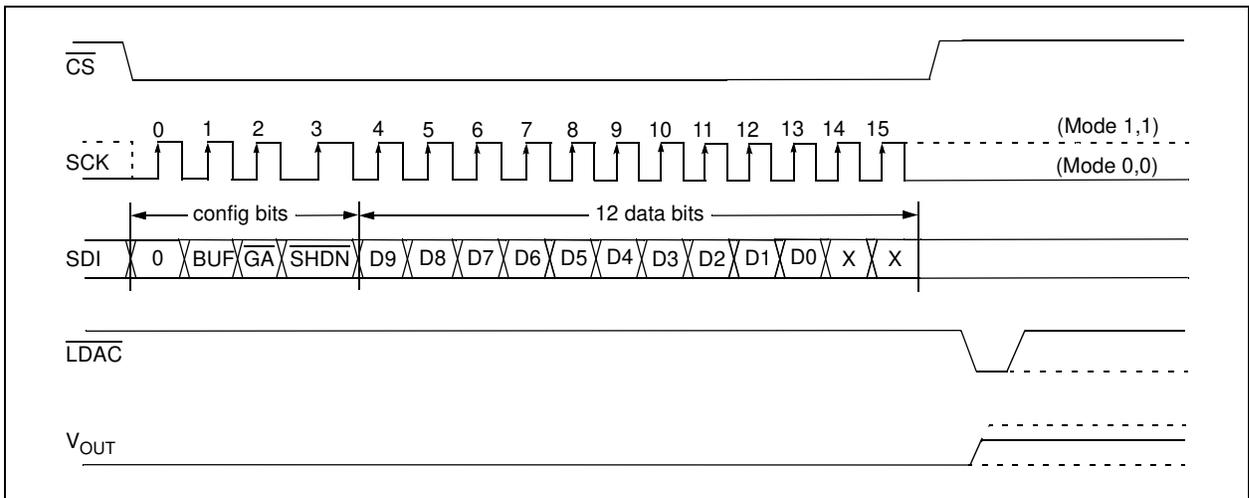


FIGURE 5-2: Write Command for MCP4911 (10-bit DAC). Note: X are don't care bits.

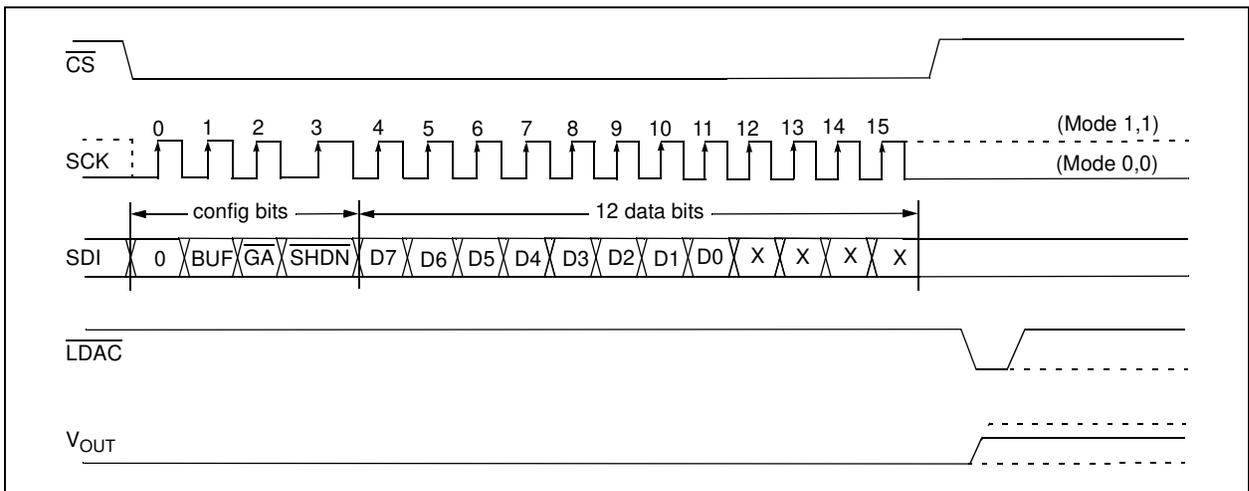


FIGURE 5-3: Write Command for MCP4901 (8-bit DAC). Note: X are don't care bits.